

30V 1MHz High Efficiency Boost LED Drive

FEATURES

- High Efficiency: Up to 90%
- 1.0MHz Constant Frequency Operation
- Integrated Power MOSFET
- Drives up to 7 Series WLEDs
- 202mV Typical V_{FB} Voltage
- Soft-start/Dimming with wide PWM Frequency Range
- UVLO, Thermal Shutdown
- Internal Current limit
- Over Voltage Protection
- Small LC Filter Components
- Minimize the External Component
- $<1\mu A$ Shutdown Current
- SOT23-6 Package

APPLICATIONS

- Camera Flash White LED
- Mobile Phone, Smart Phone LED Backlight
- PDA LED Back light
- Digital Still Cameras
- Camcorder

GENERAL DESCRIPTION

The TP7200S6 is a step-up converter designed for driving up to 7 series white LED's from a single cell Lithium Ion battery. The TP7200S6 uses current mode, fixed frequency architecture to regulate an LED current, which is measured through an external current sense resistor. Its low 202mV typical feedback voltage with EN/PWM Pulling High reduces power loss and improves efficiency.

The TP7200S6 includes under-voltage lockout, current limiting and thermal overload protection preventing damage in the event of an output overload.

Optimized operation frequency can meet the requirement of small LC filters value and low operation current with high efficiency. Internal soft start function can reduce the inrush current. SOT23-6 package type provides the best solution for PCB space saving and total BOM cost.

TYPICAL APPLICATIONS

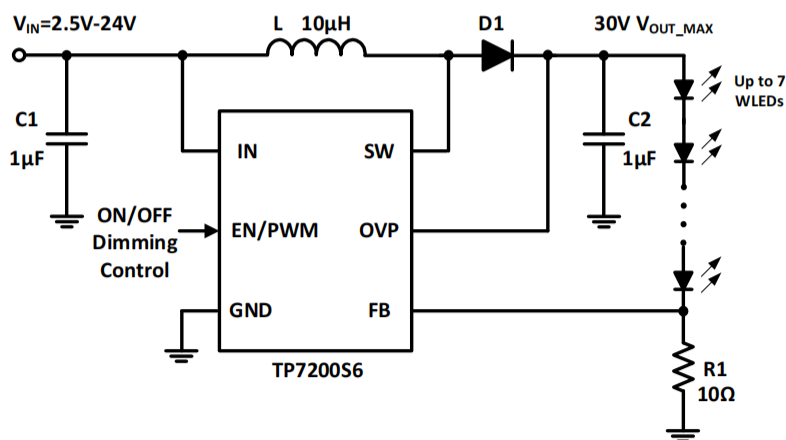
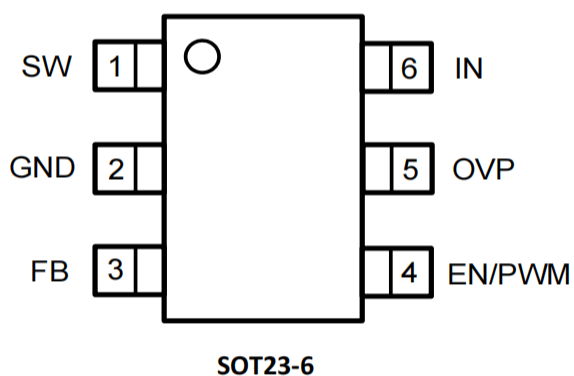


Figure 1. Basic Application Circuit

ABSOLUTE MAXIMUM RATINGS (Note1)

Parameter	Value	Unit
Input Supply voltage	-0.3 ~ 25	V
SW, OVP voltage	-0.3 ~ 35	V
EN/PWM voltage	-0.3 ~ 25	V
Junction Temperature (Note2)	160	°C
Power Dissipation	600	mW
Storage Temperature Range	-65~150	°C
Lead Temperature (Soldering, 10s)	260	°C

PACKAGE/ORDER INFORMATION



PIN DESCRIPTIONS

Pin	Name	Function
1	SW	Power Switch Output. It is the switch node connection to inductor.
2	GND	Ground Pin
3	FB	Output Voltage Feedback Pin.
4	EN	Chip Enable Pin. Drive EN above 1.5V to turn on the part. Drive EN below 0.4V to turn it off. Do not leave EN floating.
5	OVP	Over Voltage Input. OV measures the output voltage for open circuit protection. Connect OV to the output at the top of the LED string.
6	IN	Power Supply Input. Must be closely decoupled to GND with a 1 μ F or greater ceramic capacitor.

ESD RATING

Items	Description	Value	Unit
V _{ESD}	Human Body Model for all pins	±2000	V

JEDEC specification JS-001

RECOMMENDED OPERATING CONDITIONS

Items	Description	Min	Max	Unit
Voltage Range	IN	2.7	24	V
T _J	Operating Junction Temperature Range	-40	125	°C

ELECTRICAL CHARACTERISTICS

$V_{IN}=V_{EN}=3.7V$, $T_A = 25^{\circ}C$, unless otherwise noted.)

Parameter	Conditions	Min	Typ	Max	Unit
Input Voltage Range		2.7		24	V
UVLO Threshold			2.4		V
Under Voltage Lockout Hysteresis			100		mV
Shutdown Current	$V_{EN}<0.4V$		0.1	1	μA
Quiescent Current	$V_{FB}=0.15V$, No switch		200	400	μA
Supply Current	$V_{FB}=0V$, Switching		0.6	1	mA
Regulated Feedback Voltage	EN/PWM is high	197	202	207	mV
Oscillation Frequency			1.0		MHz
ON Resistance of NMOS			0.2		Ω
Peak Current Limit	$V_{IN}=4.2V$, Duty cycle=50%	1.2	1.4		A
EN High-Level Input Voltage		1.5			V
EN Low-Level Input Voltage				0.4	V
EN Leakage Current			± 0.01	± 1.0	μA
OVP Threshold	Vov Rising		30		V
OVP Hysteresis			4.5		V

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2: T_J is calculated from the ambient temperature T_A and power dissipation PD according to the following formula: $T_J = T_A + (PD) \times \theta_{JA}$.

OPERATION

The TP7200S6 uses a constant frequency, peak current mode boost regulator architecture to regulate the current of white LEDs series string. At the start of each oscillator cycle the NMOSFET is turned on through the control circuitry. To prevent sub-harmonic oscillations at duty cycles greater than 50 percent, a stabilizing ramp is added to the output of the current sense amplifier and the result is fed into the positive input of the PWM comparator. When this voltage equals the output voltage of the error amplifier, the power MOSFET is turned off. The voltage at the output of the error amplifier is an amplified version of the difference between the reference voltage and the feedback voltage. In this way the peak current level keeps the output in regulation. If the feedback voltage starts to drop, the output of the error amplifier increases. This results in more current flowing through the power MOSFET, thus increasing the power delivered to the output. When EN/PWM is connected to high logic, the inner reference of error amplifier input is 202mV. If PWM signal is supplied on EN/PWM pin, the inner reference of error amplifier input is equal to the value of 202mV multiplying duty of PWM signal.

TP7200S6 also has OVP function for LED strings open circuit protection. When OVP voltage is higher than 30V typically, device will stop switching, and the output voltage will fall down. When OVP voltage drops lower than 25.5V typically, the device will restart to switching operation.

FUNCTIONAL BLOCK DIAGRAM

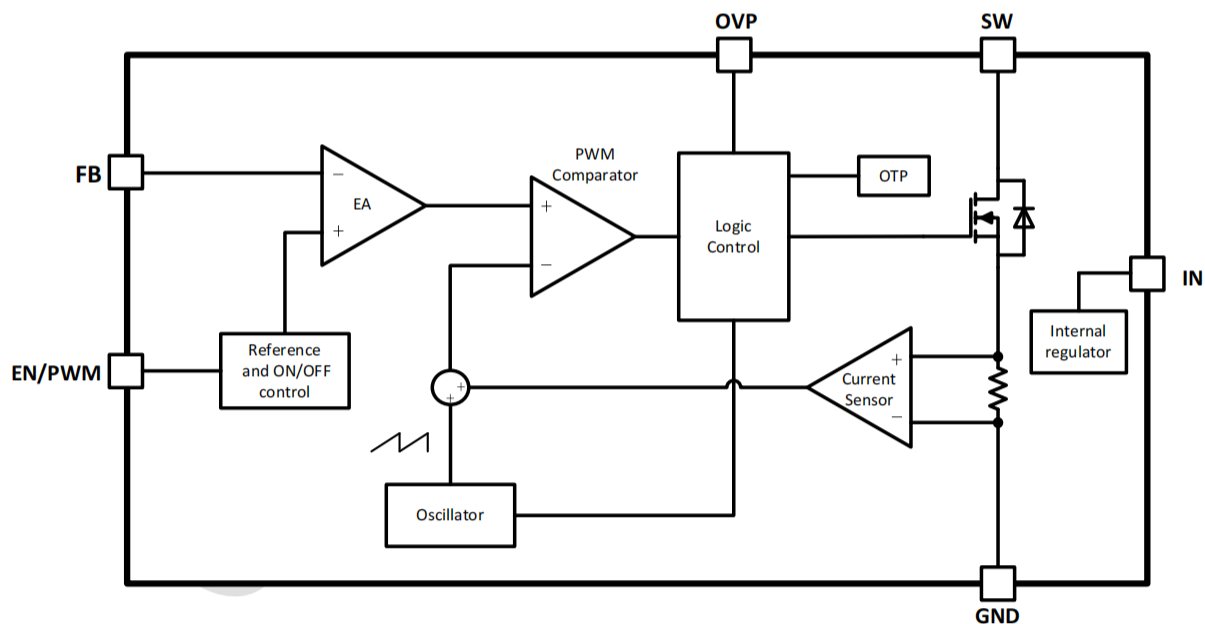


Figure 2. TP7200S6 Block Diagram

APPLICATION INFORMATION

Setting the LED current

The LED current is controlled by the feedback resistor, R1, in Figure 1. The current through the LEDs is given by the equation $202\text{mV}/R1$. Table 1 shows the selection of resistors for a given LED current.

Table 1— I_{LED} vs. R1

I_{LED} (mA)	R1 (Ω)
1	202
5	40.2
10	20
15	13.3
20	10

Dimming Control

a. Using a PWM Signal to EN Pin

For controlling the LED brightness, the TP7200S6 can perform the dimming control by applying a PWM signal to EN pin. The internal soft start and the wide range dimming frequency can eliminate inrush current and audio noise when dimming. The average LED current is proportional to the PWM signal duty cycle. The magnitude of the PWM signal should be higher than the maximum enable voltage of EN pin, in order to let the dimming control performs correctly for preventing the flicker issue, the suggested PWM frequency is $\geq 1\text{kHz}$ or $\leq 200\text{Hz}$.

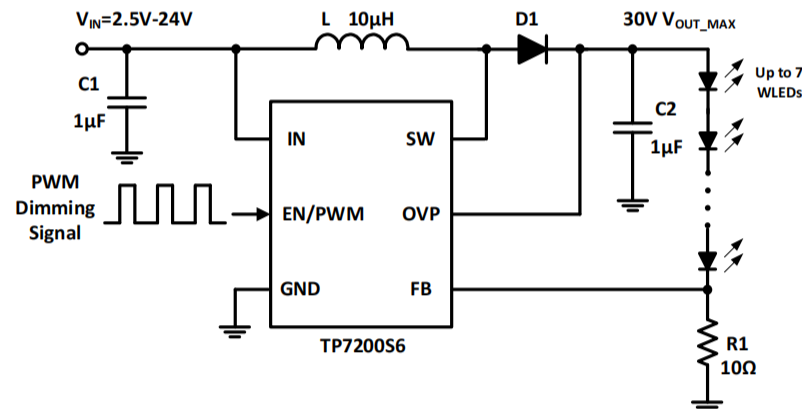


Figure 3. Dimming Control by a PWM Signal to EN Pin

b. Using a DC Voltage

Using a variable DC voltage to adjust the brightness is a popular method in some applications. According to the Superposition Theorem, as the DC voltage increases, the voltage contributed to V_{FB} increases and the voltage drop on R2 decreases, i.e. the LED current decreases. Please refer to the Figure 4, the LED current is given by the below equation:

$$I_{\text{LED}} = \frac{0.202}{\frac{R1 \cdot R2}{R1 + R2}} - \frac{V_{\text{DC}}}{R2}$$

Equation 1.

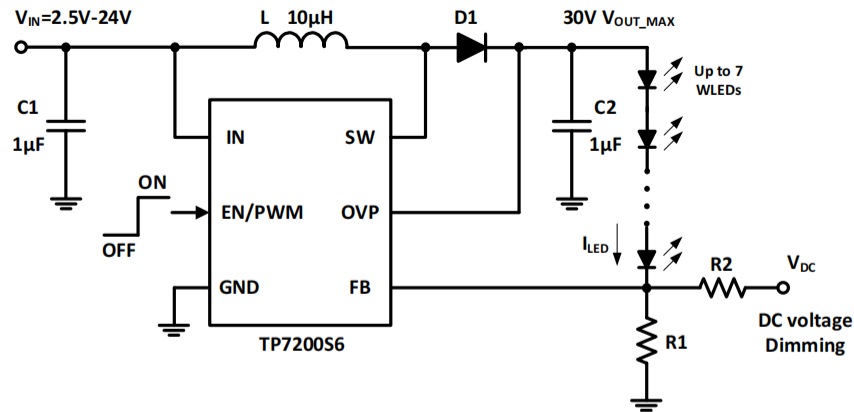


Figure 4. Dimming Control by a DC Voltage

c. Using a Filtered PWM signal

Another common application is using a filtered PWM signal as an adjustable DC voltage for LED dimming control. A filtered PWM signal acts as the DC voltage to regulate the output current. Please refer to the Figure 5, the LED current is given by the below equation:

$$I_{LED} = \frac{0.202}{\frac{R1 \cdot R2}{R1 + R2}} - \frac{\text{Duty} \cdot V_D}{R2} \quad \text{Equation 2.}$$

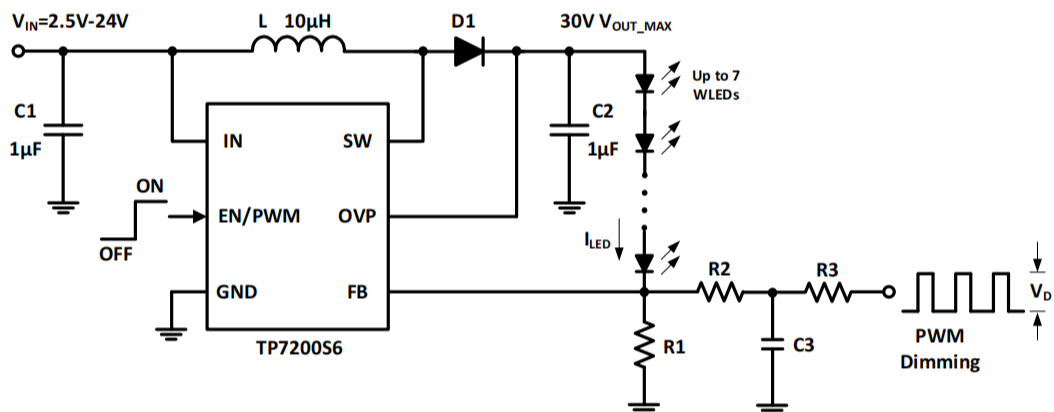


Figure 5. Dimming Control by a Filtered PWM Signal

Inductor Selection

The recommended value of inductor is 4.7 to 22μH. Small size and better efficiency are the major concerns for portable device, such as TP7200S6 used for mobile phone. The inductor should have low core loss at 1.0MHz and low DCR for better efficiency. To avoid inductor saturation current rating should be considered.

Capacitor Selection

Input and output ceramic capacitors of 1μF are recommended for TP7200S6 applications. For better voltage filtering, ceramic capacitors with low ESR are recommended. X5R and X7R types are suitable because of their wider voltage and temperature ranges.

Diode Selection

Schottky diode is a good choice for TP7200S6 because of its low forward voltage drop and fast reverse recovery. Using Schottky diode can get better efficiency. The high speed rectification is also a good

characteristic of Schottky diode for high switching frequency. Current rating of the diode must meet the root mean square of the peak current and output average current multiplication as following:

$$I_D(\text{RMS}) \approx \sqrt{I_{\text{OUT}} \times I_{\text{PEAK}}}$$

The diode's reverse breakdown voltage should be larger than the output voltage.

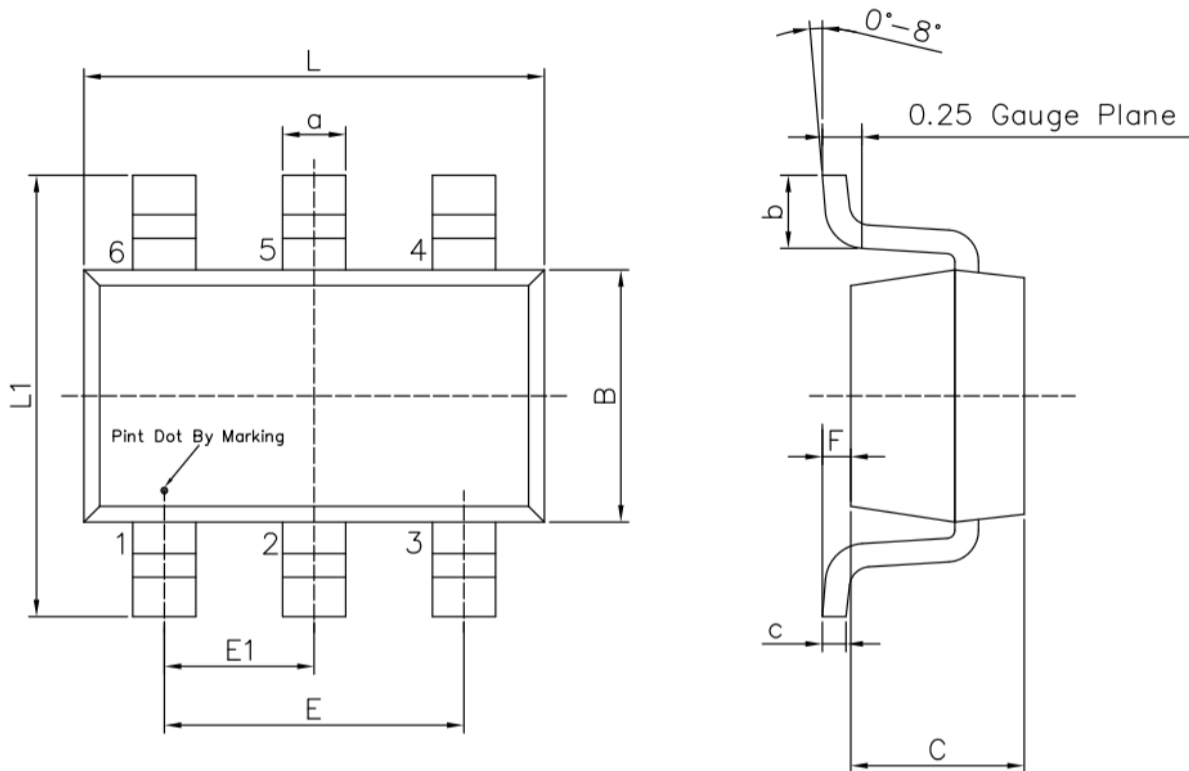
Layout Consideration

For best performance of the TP7200S6, the following guide lines must be strictly followed.

1. Input and Output capacitors should be placed close to the IC and connected to ground plane to reduce noise coupling.
2. The GND and Exposed Pad should be connected to a strong ground plane for heat sinking and noise protection.
3. Keep the main current traces as possible as short and wide.
4. SW node of DC-DC converter is with high frequency voltage swing. It should be kept at a small area.
5. Place the feedback components as close as possible to the IC and keep away from the noisy devices.

PACKAGE INFORMATION

SOT23-6



Unit: mm

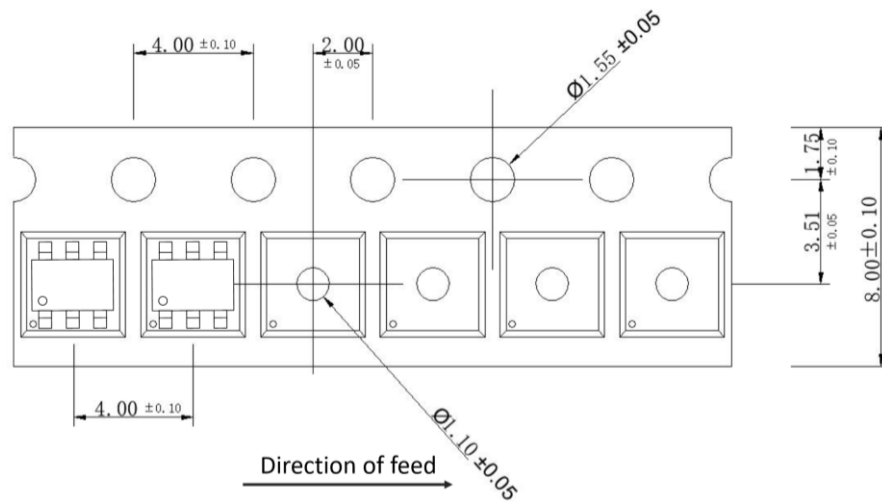
Symbol	Dimensions In Millimeters		Symbol	Dimensions In Millimeters	
	Min	Max		Min	Max
L	2.82	3.02	E1	0.85	1.05
B	1.50	1.70	a	0.35	0.50
C	0.90	1.30	c	0.10	0.20
L1	2.60	3.00	b	0.35	0.55
E	1.80	2.00	F	0	0.15

Note:

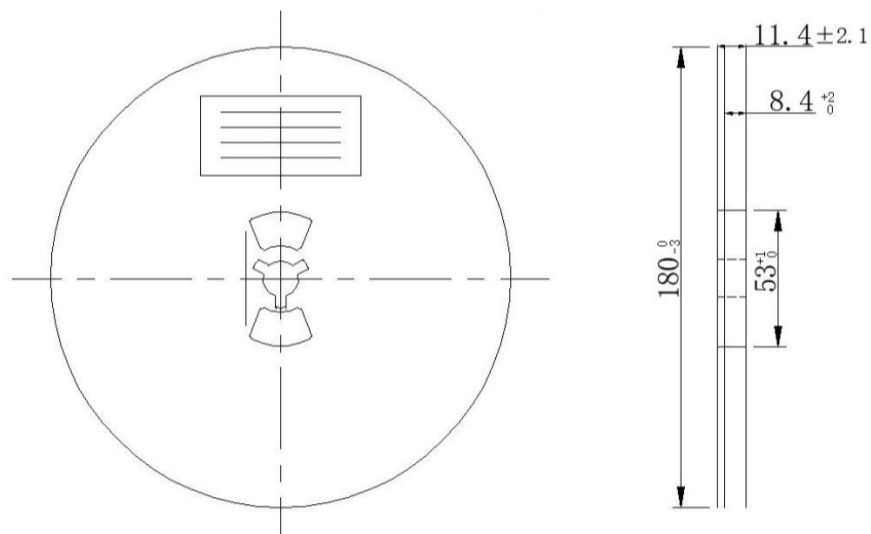
- 1) All dimensions are in millimeters.
- 2) Package length does not include mold flash, protrusion or gate burr.
- 3) Package width does not include inter lead flash or protrusion.
- 4) Lead popularity (bottom of leads after forming) shall be 0.10 millimeters max.
- 5) Pin 1 is lower left pin when reading top mark from left to right.

TAPE AND REEL INFORMATION

TAPE DIMENSIONS: SOT23-6



REEL DIMENSIONS: SOT23-6



Note:

- 1) All Dimensions are in Millimeter
- 2) Quantity of Units per Reel is 3000
- 3) MSL level is level 3.